EE 447 VLSI
Design

## Lecture 5:

Logical Effort

## Outline

- Introduction
- Delay in a Logic Gate
- Multistage Logic Networks
- Choosing the Best Number of Stages
- Example
- Summary


## Introduction

- Chip designers face a bewildering array of choices
- What is the best circuit topology for a function?
- How many stages of logic give least delay?
- How wide should the transistors be?
- Logical effort is a method to make these decisions
- Uses a simple model of delay
- Allows back-of-the-envelope calculations
- Helps make rapid comparisons between alternatives
- Emphasizes remarkable symmetries


## Example

- Ben Bitdiddle is the memory designer for the Motoroil 68W86, an embedded automotive processor. Help ${ }_{\text {Al }}$ Ben design the decoder for a register file.
- Decoder specifications:
- 16 word register file
- Each word is 32 bits wide
- Each bit presents load of 3 unit-sized transistors
- True and complementary address inputs A[3:0]
- Each input may drive 10 unit-sized transistors
- Ben needs to decide:
- How many stages to use?
- How large should each gate be?
- How fast can decoder operate?


## Delay in a Logic Gate

Express delays in process-independent unit

$$
d=\frac{d_{a b s}}{\tau}
$$

$$
\tau=3 \mathrm{RC}
$$

$\approx 12 \mathrm{ps}$ in 180 nm process
40 ps in $0.6 \mu \mathrm{~m}$ process

## Delay in a Logic Gate

- Express delays in process-independent unit

$$
d=\frac{d_{a b s}}{\tau}
$$

- Delay has two components

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d=f+p
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- Again has two components


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- Effort delay $f=g h$ (a.k.a. stage effort)
- Again has two components
- g: logical effort
- Measures relative ability of gate to deliver current
- $g \equiv 1$ for inverter


## Delay in a Logic Gate

- Express delays in process-independent unit

$$
d=\frac{d_{a b s}}{\tau}
$$

- Delay has two components $d=f+p$
- Effort delay $f=g h$ (a.k.a. stage effort)
- Again has two components
- h: electrical effort $=\mathrm{C}_{\text {out }} / \mathrm{C}_{\text {in }}$
- Ratio of output to input capacitance
- Sometimes called fanout


## Delay in a Logic Gate

Express delays in process-independent unit

$$
d=\frac{d_{a b s}}{\tau}
$$

- Delay has two components
$d=f+p$
- Parasitic delay $p$
- Represents delay of gate driving no load
- Set by internal parasitic capacitance


## Delay Plots

$$
\begin{aligned}
d & =f+p \\
& =g h+p
\end{aligned}
$$



## Delay Plots

$\square$ What about NOR2?


## Computing Logical Effort

- DEF: Logical effort is the ratio of the input capacitance of a gate to the input capacitance of an inverter delivering the same output current.
- Measure from delay vs. fanout plots
- Or estimate by counting transistor ${ }^{2}$ widths


$$
\begin{aligned}
& C_{\text {in }}=3 \\
& g=3 / 3
\end{aligned}
$$

$C_{i n}=4$
$g=4 / 3$
$C_{\text {in }}=5$
$g=5 / 3$

## Catalog of Gates

- Logical effort of common gates

| Gate type | Number of inputs |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | 1 | 2 | 3 | 4 | n |
| Inverter | 1 |  |  |  |  |
| NAND |  | 4/3 | 5/3 | 6/3 | $(\mathrm{n}+2) / 3$ |
| NOR |  | 5/3 | 7/3 | 9/3 | $(2 n+1) / 3$ |
| Tristate / mux | 2 | 2 | 2 | 2 | 2 |
| XOR, XNOR |  | $\begin{array}{r} 4,4 \\ \hline \end{array}$ | $6,12,$ $4{ }^{4} \text { evesinse }$ | $\begin{aligned} & 8,16,16, \\ & 8 \end{aligned}$ |  |

## Catalog of Gates

- Parasitic delay of common gates
- In multiples of $\mathrm{p}_{\text {inv }}(\approx 1)$

| Gate type | Number of inputs |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  | 1 | 2 | 3 | 4 | n |
| Inverter | 1 |  |  |  |  |
| NAND |  | 2 | 3 | 4 | n |
| NOR |  | 2 | 3 | 4 | n |
| Tristate $/$ <br> mux | 2 | 4 | 6 | 8 | 2 n |
| XOR, XNOR |  | 4 | 6 | 8 |  |

## Example: Ring Oscillator

- Estimate the frequency of an N -stage ring oscillator

$\begin{array}{ll}\text { Logical Effort: } & \mathrm{g}= \\ \text { Electrical Effort: } & \mathrm{h}= \\ \text { Parasitic Delay: } & \mathrm{p}=\end{array}$
Stage Delay: $\mathrm{d}=$
Frequency: $f_{\text {osc }}=$


## Example: Ring Oscillator

- Estimate the frequency of an N -stage ring oscillator

$\begin{array}{ll}\text { Logical Effort: } & g=1 \\ \text { Electrical Effort: } & h=1 \\ \text { Parasitic Delay: } & p=1\end{array}$
31 stage ring oscillator in $0.6 \mu \mathrm{~m}$ process has frequency of $\sim 200 \mathrm{MHz}$
Stage Delay: $\mathrm{d}=2$
Frequency: $f_{\text {osc }}=1 /\left(2^{*} N^{*} d\right)=1 / 4 N$


## Example: FO4 Inverter

- Estimate the delay of a fanout-of-4 (FO4) inverter


Logical Effort: $\quad \mathrm{g}=$
Electrical Effort: h =
Parasitic Delay: p=
Stage Delay: d=

## Example: FO4 Inverter

- Estimate the delay of a fanout-of-4 (FO4) inverter


Logical Effort: $\quad g=1$
Electrical Effort: $\mathrm{h}=4$
Parasitic Delay: p=1
Stage Delay: d=5
The FO4 delay is about 200 ps in $0.6 \mu \mathrm{~m}$ process

60 ps in a 180 nm process $\mathrm{f} / 3 \mathrm{~ns}$ in an $f \mu \mathrm{~m}$ process

## Multistage Logic Networks

- Logical effort generalizes to multistage networks
- Path Logical Effort $\quad G=\prod g_{i}$
- Path Electrical Effort $H=\frac{C_{\text {out-path }}}{C_{\text {in-path }}}$
- Path Effort

$$
F=\prod f_{i}=\prod g_{i} h_{i}
$$



## Multistage Logic Networks

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$$

Can we write $\mathrm{F}=\mathrm{GH}$ ?

## Paths that Branch

- No! Consider paths that branch:



## Paths that Branch

No! Consider paths that branch:

$$
\begin{aligned}
& \mathrm{G}=1 \\
& \mathrm{H}=90 / 5=18 \\
& \mathrm{GH}=18 \\
& \mathrm{~h}_{1}=(15+15) / 5=6 \\
& \mathrm{~h}_{2}=90 / 15=6 \\
& \mathrm{~F}=\mathrm{g}_{1} \mathrm{~g}_{2} \mathrm{~h}_{1} \mathrm{~h}_{2}=36=2 \mathrm{GH}
\end{aligned}
$$

## Branching Effort

- Introduce branching effort
- Accounts for branching between stages in path

$$
b=\frac{C_{\text {on path }}+C_{\text {off path }}}{C_{\text {on path }}}
$$

$$
B=\prod b_{i}
$$

Note:

$$
\prod^{h_{1}=B H}
$$

- Now we compute the path effort
$\square F=G B H$


## Multistage Delays

- Path Effort Delay

$$
D_{F}=\sum f_{i}
$$

- Path Parasitic Delay $P=\sum p_{i}$
- Path Delay

$$
D=\sum d_{i}=D_{F}+P
$$

## Designing Fast Circuits

$$
D=\sum d_{i}=D_{F}+P
$$

- Delay is smallest when each stage bears same effort

$$
\hat{f}=g_{i} h_{i}=F^{\frac{1}{N}}
$$

- Thus minimum delay of $N$ stage path is

$$
D=N F^{\frac{1}{N}}+P
$$

- This is a key result of logical effort
- Find fastest possible delay
- Doesn't require calculating gate sizes


## Gate Sizes

- How wide should the gates be for least delay?

$$
\begin{aligned}
\hat{f} & =g h=g \frac{C_{o u t}}{C_{\text {in }}} \\
& \Rightarrow C_{i n_{i}}=\frac{g_{i} C_{o u t_{i}}}{\hat{f}}
\end{aligned}
$$

- Working backward, apply capacitance transformation to find input capacitance of each gate given load it drives.
- Check work by verifying input cap spec is met.


## Example: 3-stage path

- Select gate sizes $x$ and $y$ for least delay from



## Example: 3-stage path



Logical Effort
G =
Electrical Effort
$\mathrm{H}=$
Branching Effort
$\mathrm{B}=$
Path Effort
$\hat{f}=$
Parasitic Delay
$\mathrm{P}=$
Delay
D =

## Example: 3-stage path



Logical Effort $G=(4 / 3)^{\star}(5 / 3) *(5 / 3)=100 / 27$
Electrical Effort $H=45 / 8$
Branching Effort
$B=3 * 2=6$
Path Effort
Best Stage Effort
$\hat{f}=\sqrt[3]{F}=\sqrt[3]{F}=125$
Parasitic Delay
$P=2+3+2=7$
Delay
$\mathrm{D}=3 * 5+7=22=4.4 \mathrm{FO} 4$

## Example: 3-stage path

- Work backward for sizes

$$
\begin{aligned}
& y= \\
& x=
\end{aligned}
$$



## Example: 3-stage path

- Work backward for sizes

$$
\begin{aligned}
& y=45 *(5 / 3) / 5=15 \\
& x=(15 * 2) *(5 / 3) / 5=10
\end{aligned}
$$



## Best Number of Stages

- How many stages should a path use?
- Minimizing number of stages is not always fastest
- Example: drive 64-bit datapath with unit inverter

D =


## Best Number of Stages

- How many stages should a path use?
- Minimizing number of stages is not always fastest
- Example: drive 64-bit datapath with unit inverter

$$
\begin{aligned}
D & =N F^{1 / N}+P \\
& =N(64)^{1 / N}+N
\end{aligned}
$$



## Derivation

- Consider adding inverters to end of path

$$
\begin{aligned}
& \text { - How many give least delay? Logic Elock: }{ }^{N-n_{1} \text { Extranverters }}
\end{aligned}
$$

$$
\begin{aligned}
& \frac{\partial D}{\partial N}=-F^{\frac{1}{N}} \ln F^{\frac{1}{N}}+F^{\frac{1}{N}}+p_{i n v}=0
\end{aligned}
$$

- Define best stage effort $\rho=F^{\frac{1}{N}}$

$$
p_{i n v}+\rho(1-\ln \rho)=0
$$

## Best Stage Effort

$p_{\text {inv }}+\rho(1-\ln \rho)=0$
has no closed-form solution
$\square$ Neglecting parasitics $\left(p_{\text {inv }}=0\right)$, we find $\rho=$ 2.718 (e)
$\square$ For $p_{\text {inv }}=1$, solve numerically for $\rho=3.59$

## Sensitivity Analysis

- How sensitive is delay to using exactly the best number of stages?

- $2.4<\rho<6$ gives delay within $15 \%$ of optimal
- We can be sloppy!
- I like $\rho=4$


## Example, Revisited

- Ben Bitdiddle is the memory designer for the Motoroil 68W86, an embedded automotive processor. Help Ben design the decoder for a register file.
- Decoder specifications:
- 16 word register file
- Each word is 32 bits wide

- Each bit presents load of 3 unit-sized transistors
- True and complementary address inputs A[3:0]
- Each input may drive 10 unit-sized transistors
- Ben needs to decide:
- How many stages to use?
- How large should each gate be?
- How fast can decoder operate?


## Number of Stages

Decoder effort is mainly electrical and branching
Electrical Effort:
$\mathrm{H}=$
Branching Effort:
$B=$

If we neglect logical effort (assume G = 1)
Path Effort:
$\mathrm{F}=$

Number of Stages:
$N=$

## Number of Stages

Decoder effort is mainly electrical and branching Electrical Effort:

$$
H=(32 * 3) / 10=9.6
$$

Branching Effort: $\quad \mathrm{B}=8$

- If we neglect logical effort (assume G=1)

Path Effort:
$\mathrm{F}=\mathrm{GBH}=76.8$

Number of Stages:
$N=\log _{4} \mathrm{~F}=3.1$

- Try a 3-stage design


## Gate Sizes \& Delay

Logical Effort: $\quad \mathrm{G}=$
Path Effort: F =
Stage Effort:
Path Delay:
Gate sizes: $\mathrm{z}=\quad \mathrm{D}=$
$y=$


## Gate Sizes \& Delay

Logical Effort: $\quad G=1 * 6 / 3 * 1=2$
Path Effort: $\quad \mathrm{F}=\mathrm{GBH}=154$
Stage Effort: $\quad \hat{f}=F_{\hat{f}}^{1 / 3}=5.36$
Path Delay: $D=3 \hat{f}+1+4+1=22.1$
Gate sizes: $z=96 * 1 / 5.36=18 \quad y=18 * 2 / 5.36=6.7$


## Comparison

- Compare many alternatives with a spreadsheet

| Design | N | $\mathbf{G}$ | $\mathbf{P}$ | $\mathbf{D}$ |
| :--- | :--- | :--- | :--- | :--- |
| NAND4-INV | 2 | 2 | 5 | 29.8 |
| NAND2-NOR2 | 2 | $20 / 9$ | 4 | 30.1 |
| INV-NAND4-INV | 3 | 2 | 6 | 22.1 |
| NAND4-INV-INV-INV | 4 | 2 | 7 | 21.1 |
| NAND2-NOR2-INV-INV | 4 | $20 / 9$ | 6 | 20.5 |
| NAND2-INV-NAND2-INV | 4 | $16 / 9$ | 6 | 19.7 |
| INV-NAND2-INV-NAND2-INV | 5 | $16 / 9$ | 7 | 20.4 |
| NAND2-INV-NAND2-INV-INV-INV | 6 | $16 / 9$ | 8 | 21.6 |

## Review of Definitions

| Term | Stage | Path |
| :---: | :---: | :---: |
| number of stages | 1 | $N$ |
| logical effort | $g$ | $G=\prod g_{i}$ |
| electrical effort | $h=\frac{C_{\text {aut }}}{C_{\text {min }}}$ | $H=\frac{C_{\text {atamemh }}}{C_{\text {mpant }}}$ |
| branching effort |  | $B=\prod b_{i}$ |
| effort | $f=g h$ | $F=G B H$ |
| effort delay | $f$ | $D_{F}=\sum f_{i}$ |
| parasitic delay | $p$ | $P=\sum p_{i}$ |
| delay | $d=f+p$ | $D=\sum d_{i}=D_{F}+P$ |

## Method of Logical Effort

1) Compute path effort

$$
F=G B H
$$

2) Estimate best number of stages ${ }^{N}=\log _{4} F$
3) Sketch path with $N$ stages

$$
D=N F^{\frac{1}{N}}+P
$$

4) Estimate least delay
5) Determine best stage effort
6) Find gate sizes

$$
C_{i n_{i}}=\frac{g_{i} C_{o u t_{i}}}{\hat{f}}
$$

## Limits of Logical Effort

- Chicken and egg problem
- Need path to compute G
- But don't know number of stages without G
- Simplistic delay model
- Neglects input rise time effects
- Interconnect
- Iteration required in designs with wire
- Maximum speed only
- Not minimum area/power for constrained delay


## Summary

- Logical effort is useful for thinking of delay in circuits
- Numeric logical effort characterizes gates
- NANDs are faster than NORs in CMOS
- Paths are fastest when effort delays are $\sim 4$
- Path delay is weakly sensitive to stages, sizes
- But using fewer stages doesn't mean faster paths
- Delay of path is about $\log _{4} \mathrm{~F} \mathrm{FO} 4$ inverter delays
- Inverters and NAND2 best for driving large caps
- Provides language for discussing fast circuits
- But requires practice to master

